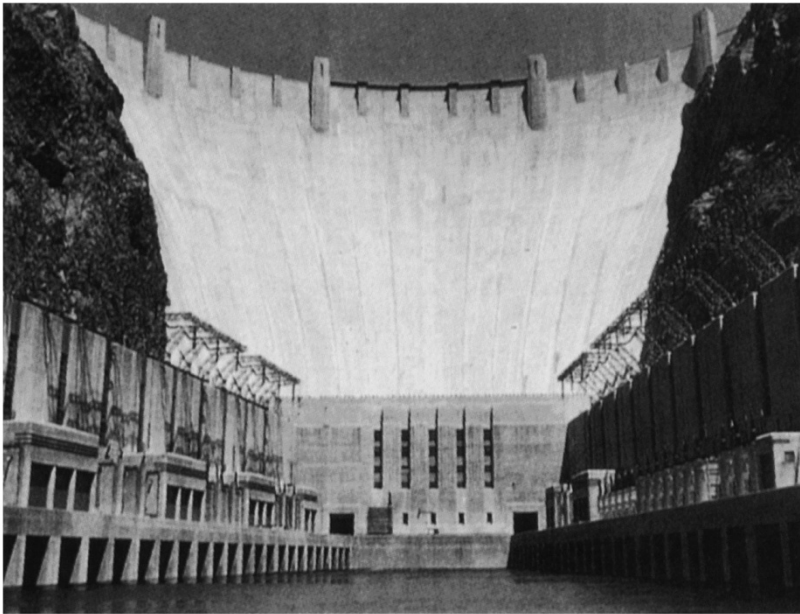
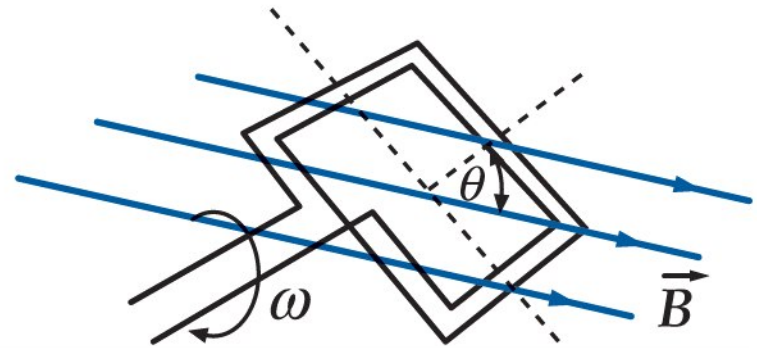
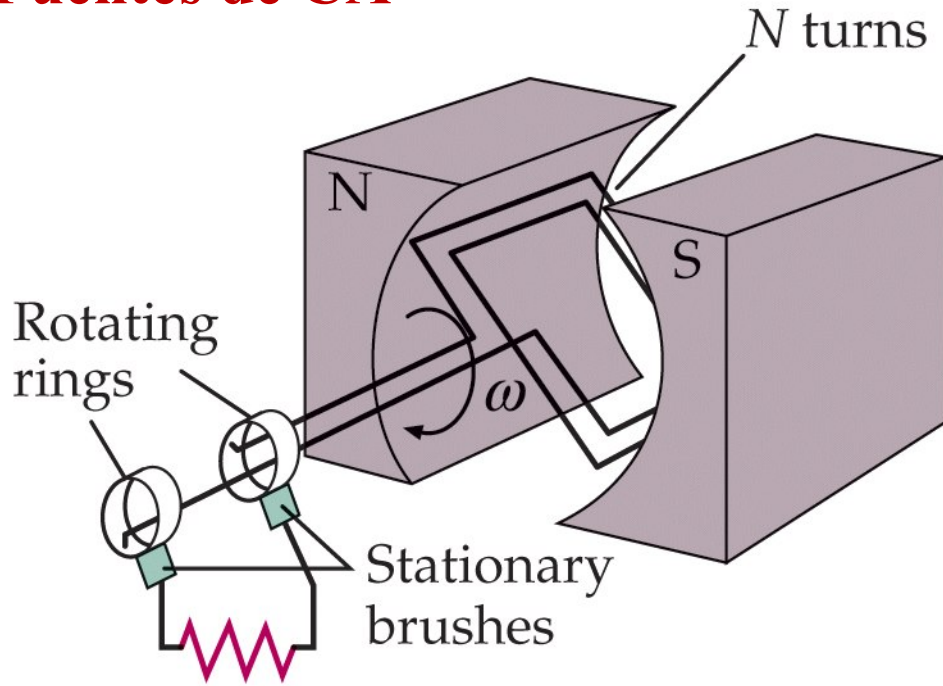


## Capítulo 6

# Circuitos de Corriente Alterna



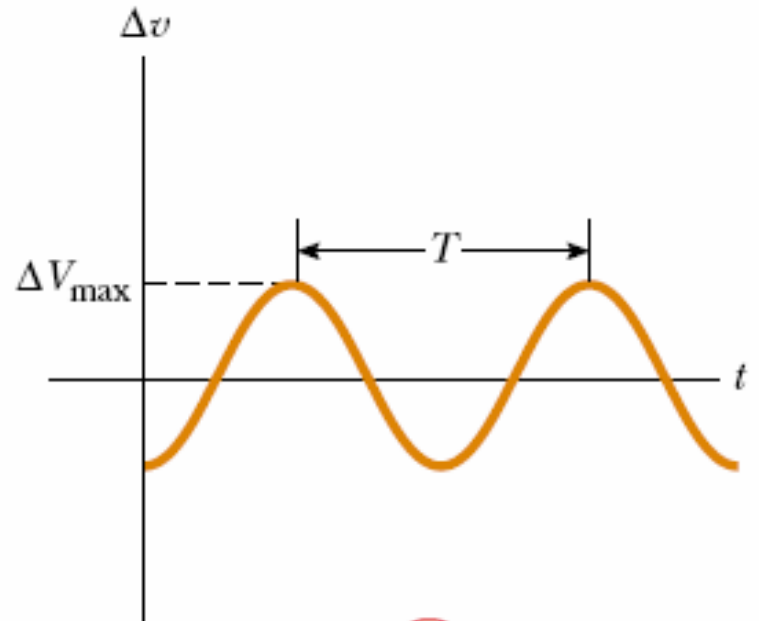
# Fuentes de CA



$$\Delta v = \Delta V_{\max} \sin \omega t$$

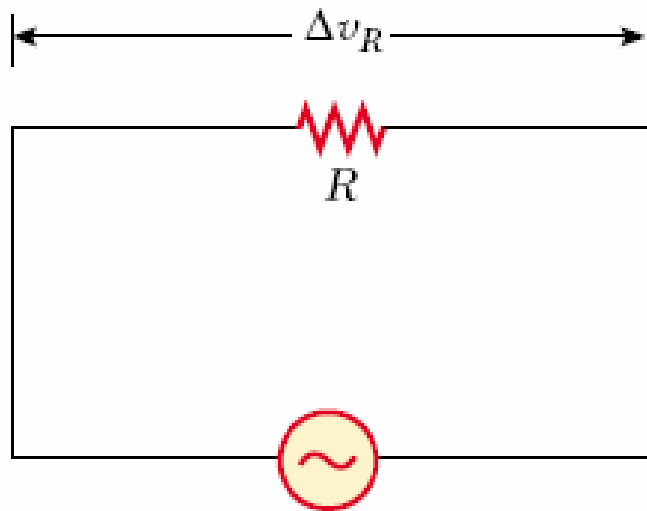
$\Delta V_{\max}$  Voltaje máximo o amplitud

$$\omega = 2\pi f = \frac{2\pi}{T} \quad \text{frecuencia angular}$$



Símbolo

## Resistores en un circuito de CA



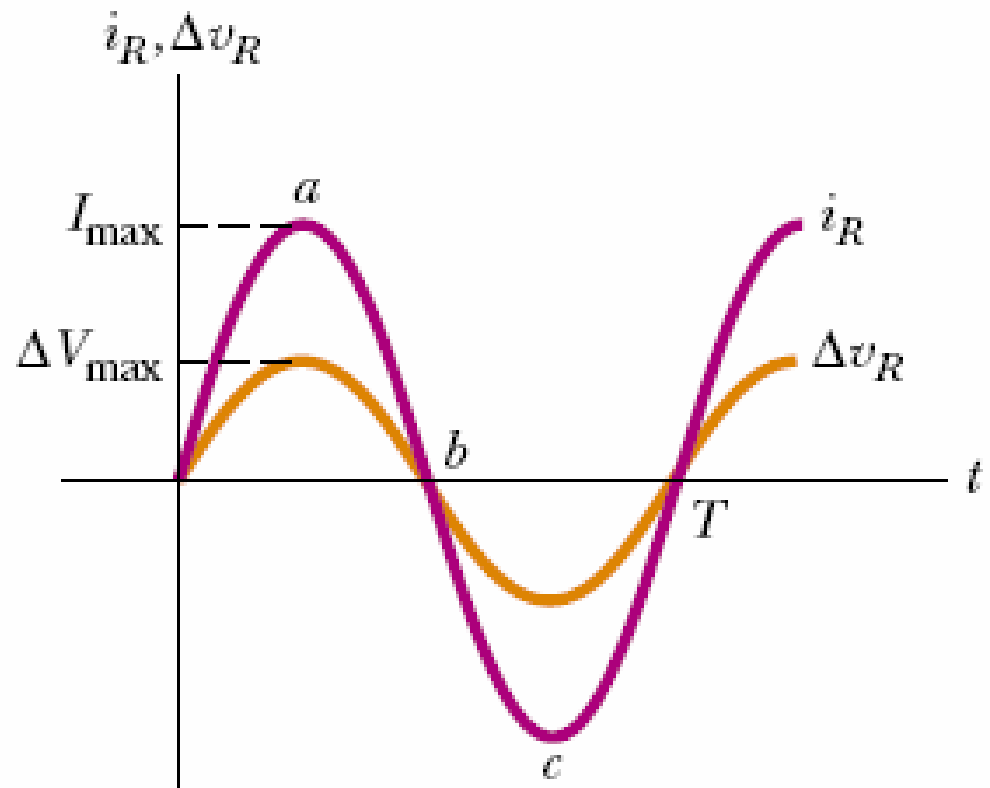
$$\Delta v = \Delta V_{\max} \sin \omega t$$

$$I_{\max} = \frac{\Delta V_{\max}}{R}$$

$$\Delta v_R = I_{\max} R \sin \omega t$$

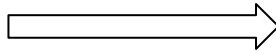
$$\Delta v = \Delta v_R = \Delta V_{\max} \sin \omega t$$

$$i_R = \frac{\Delta v_R}{R} = \frac{\Delta V_{\max}}{R} \sin \omega t = I_{\max} \sin \omega t$$

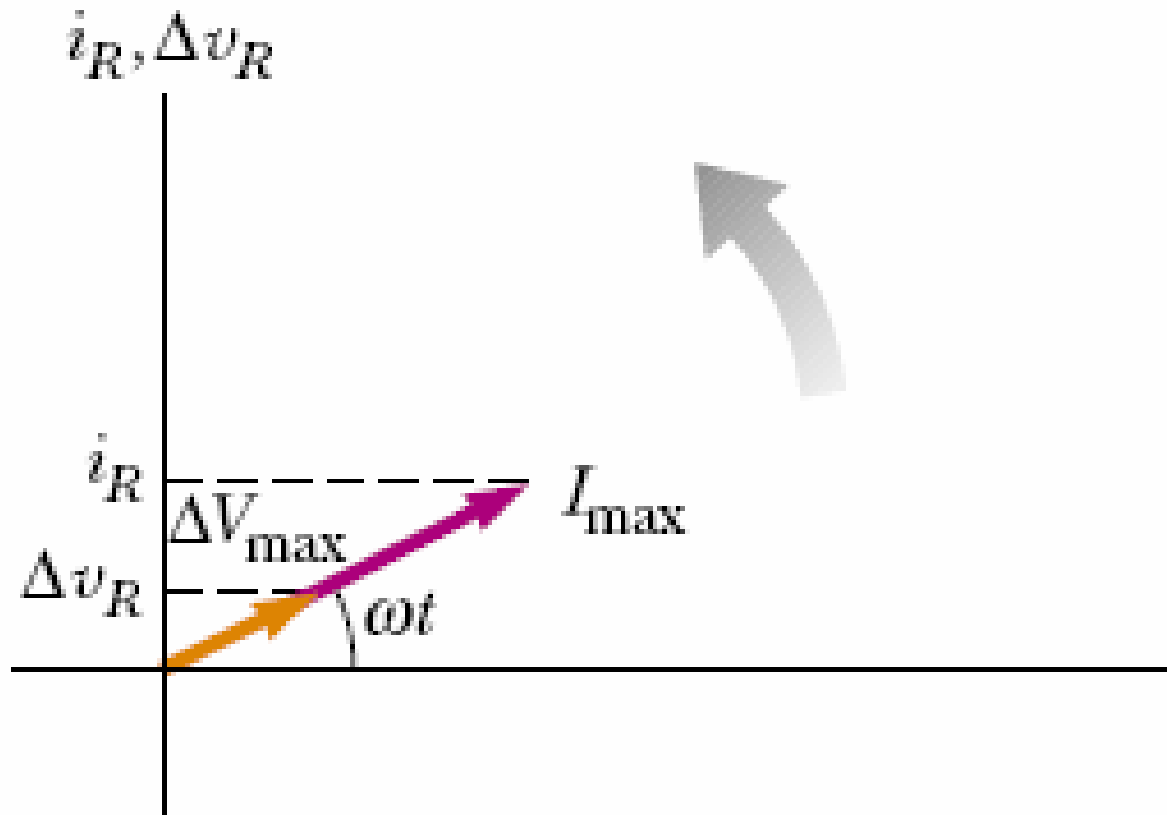


Corriente y voltaje alcanzan valores máximos en el mismo instante de tiempo: se dice que están **en fase**

Se representan con  
vectores rotatorios



Fasores

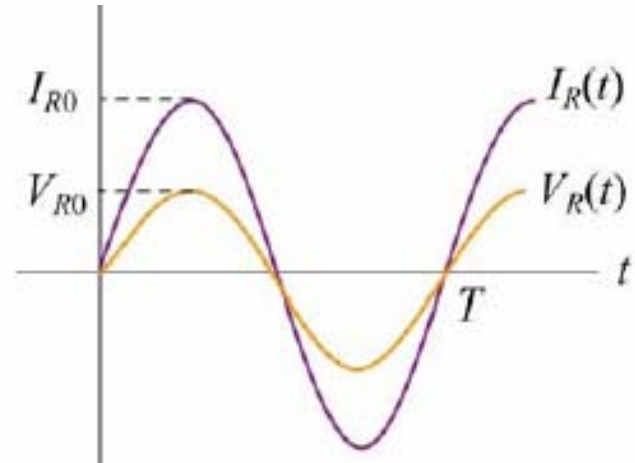


## La potencia disipada en el resistor (calor Joule)

$$I_R(t) = \frac{V_R(t)}{R} = \frac{V_{R0} \sin \omega t}{R} = I_{R0} \sin \omega t$$

$$P = I_{R0}^2 R \quad ?$$

en CA  $\longrightarrow$   $P_{\text{media}}$



$$\langle I_R(t) \rangle = \frac{1}{T} \int_0^T I_R(t) dt = \frac{1}{T} \int_0^T I_{R0} \sin \omega t dt = \frac{I_{R0}}{T} \int_0^T \sin \frac{2\pi t}{T} dt = 0$$

porque  $\langle \sin \omega t \rangle = \frac{1}{T} \int_0^T \sin \omega t dt = 0$

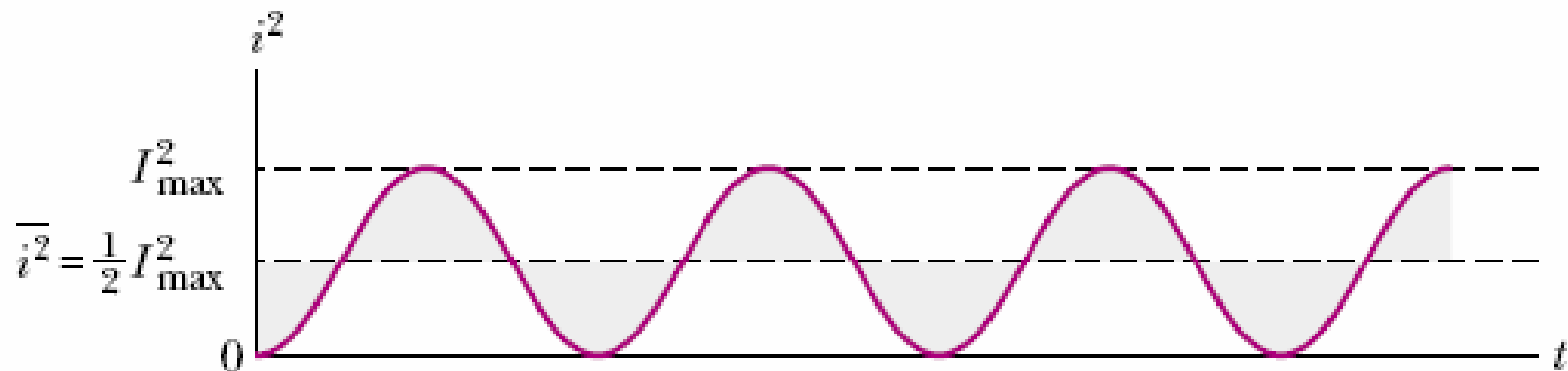
$$\langle \cos \omega t \rangle = \frac{1}{T} \int_0^T \cos \omega t \, dt = 0$$

$$\langle \sin \omega t \cos \omega t \rangle = \frac{1}{T} \int_0^T \sin \omega t \cos \omega t \, dt = 0$$

$$\langle \sin^2 \omega t \rangle = \frac{1}{T} \int_0^T \sin^2 \omega t \, dt = \frac{1}{T} \int_0^T \sin^2 \left( \frac{2\pi t}{T} \right) dt = \frac{1}{2}$$

$$\langle \cos^2 \omega t \rangle = \frac{1}{T} \int_0^T \cos^2 \omega t \, dt = \frac{1}{T} \int_0^T \cos^2 \left( \frac{2\pi t}{T} \right) dt = \frac{1}{2}$$

$$\langle I_R^2(t) \rangle = \frac{1}{T} \int_0^T I_R^2(t) dt = \frac{1}{T} \int_0^T I_{R0}^2 \sin^2 \omega t \, dt = I_{R0}^2 \frac{1}{T} \int_0^T \sin^2 \left( \frac{2\pi t}{T} \right) dt = \frac{1}{2} I_{R0}^2$$



**Es conveniente definir la corriente cuadrática media  $I_{\text{rms}}$  (rms: root-mean-square) también denominada corriente eficaz**

$$I_{\text{rms}} = \sqrt{\langle I_R^2(t) \rangle} = \frac{I_{R0}}{\sqrt{2}}$$

**En forma similar para el voltaje:**

$$V_{\text{rms}} = \sqrt{\langle V_R^2(t) \rangle} = \frac{V_{R0}}{\sqrt{2}}$$

**La potencia instantánea disipada en el resistor es:**

$$P_R(t) = I_R(t) V_R(t) = I_R^2(t) R$$

**Con lo cual, la potencia media sobre un periodo es:**

$$\langle P_R(t) \rangle = \langle I_R^2(t) R \rangle = \frac{1}{2} I_{R0}^2 R = I_{\text{rms}}^2 R = I_{\text{rms}} V_{\text{rms}} = \frac{V_{\text{rms}}^2}{R}$$

$$I_{\text{rms}} = \frac{I_{\text{max}}}{\sqrt{2}} = 0.707I_{\text{max}}$$

$$\Delta V_{\text{rms}} = \frac{\Delta V_{\text{max}}}{\sqrt{2}} = 0.707 \Delta V_{\text{max}}$$

**Son valores ficticios en CA que producen la misma potencia que en un circuito de CC**

**En la línea domiciliaria:**

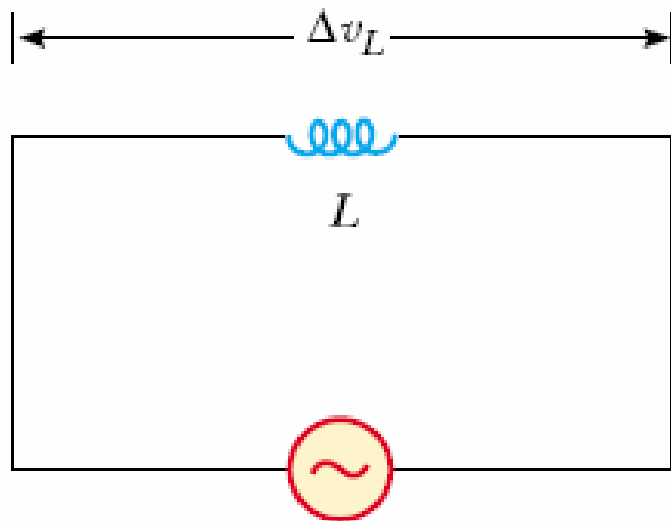
$$V_{\text{rms}} = 220 \text{ V}$$

$$V_{\text{max}} = 312 \text{ V}$$

**Amperímetros y voltímetros miden valores eficaces.**



## Inductores en un circuito de CA



$$\Delta v = \Delta V_{\max} \sin \omega t$$

$$\Delta v + \Delta v_L = 0$$

$$\Delta v - L \frac{di}{dt} = 0$$

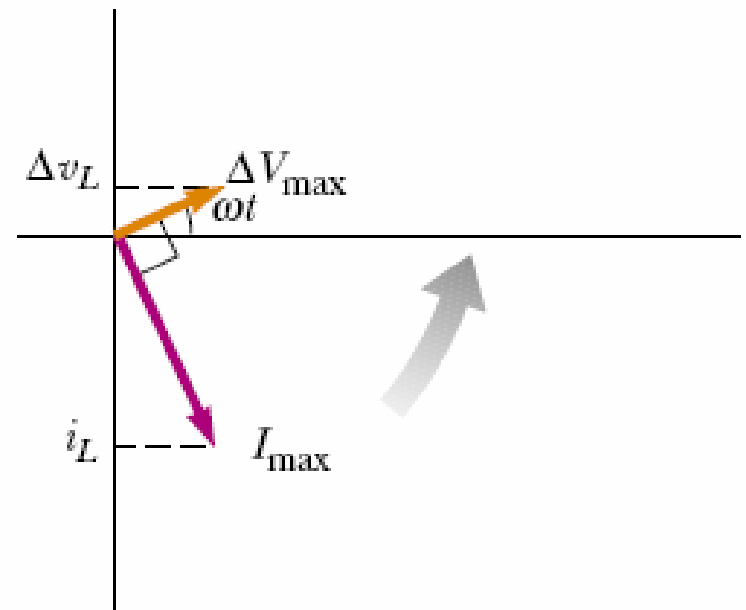
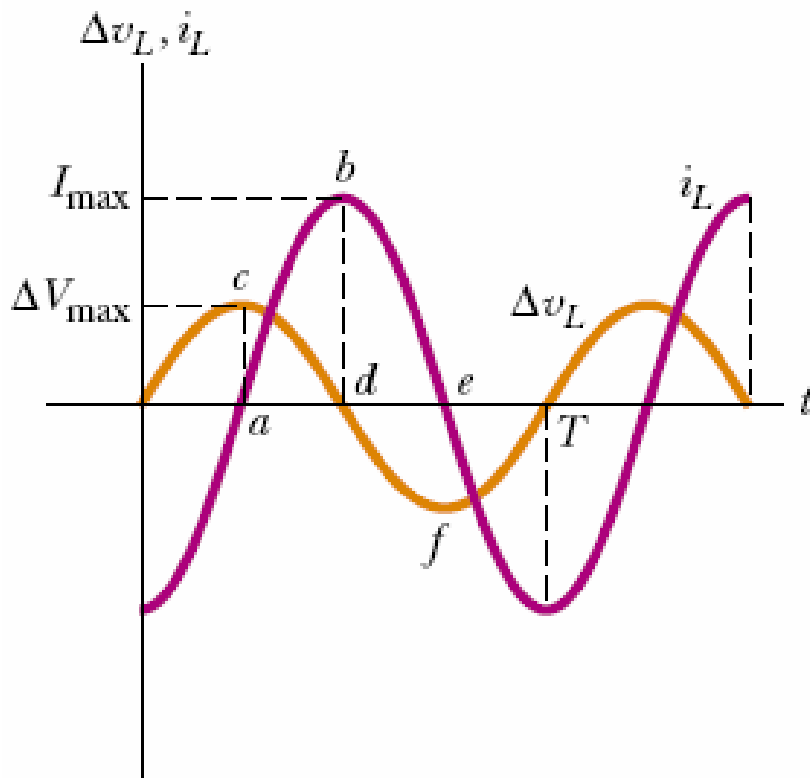
$$\Delta v = L \frac{di}{dt} = \Delta V_{\max} \sin \omega t$$

$$di = \frac{\Delta V_{\max}}{L} \sin \omega t dt$$

$$i_L = \frac{\Delta V_{\max}}{L} \int \sin \omega t dt = -\frac{\Delta V_{\max}}{\omega L} \cos \omega t$$

**Usando:**  $\cos \omega t = -\sin(\omega t - \pi/2)$

$$i_L = \frac{\Delta V_{\max}}{\omega L} \sin\left(\omega t - \frac{\pi}{2}\right)$$



La corriente está **retrasada** respecto al voltaje en  $\pi/2$

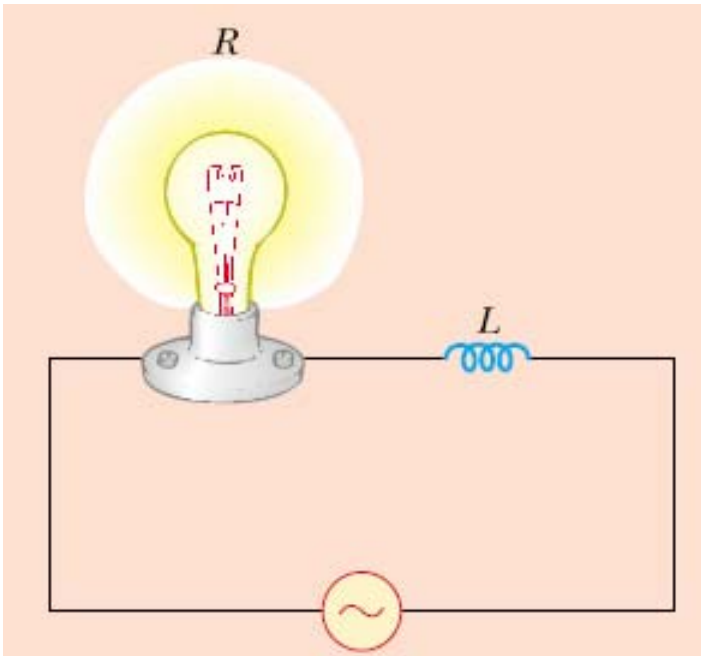
$$I_{\max} = \frac{\Delta V_{\max}}{\omega L}$$

$$X_L \equiv \omega L$$

**reactancia  
inductiva**

$$I_{\max} = \frac{\Delta V_{\max}}{X_L}$$

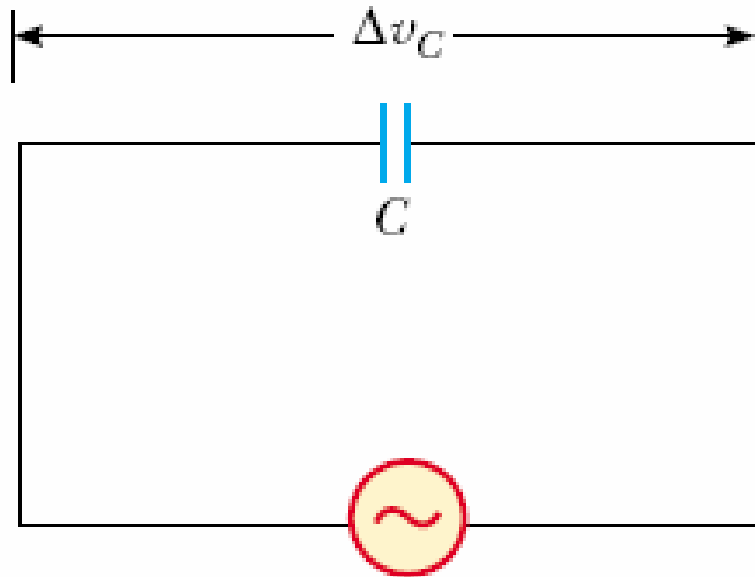
$$[ X_L ] = \Omega$$



**Si la frecuencia de la fuente es variable y la amplitud de  $V$  constante. La lámpara brilla más intensamente a:**

- i) altas frecuencias**
- ii) bajas frecuencias**
- iii) igual para todas**

## Capacitores en un circuito CA



$$\Delta v = \Delta V_{\max} \sin \omega t$$

$$\Delta v + \Delta v_C = 0,$$

$$\Delta v = \Delta v_C = \Delta V_{\max} \sin \omega t$$

$$C = q / \Delta v_C$$

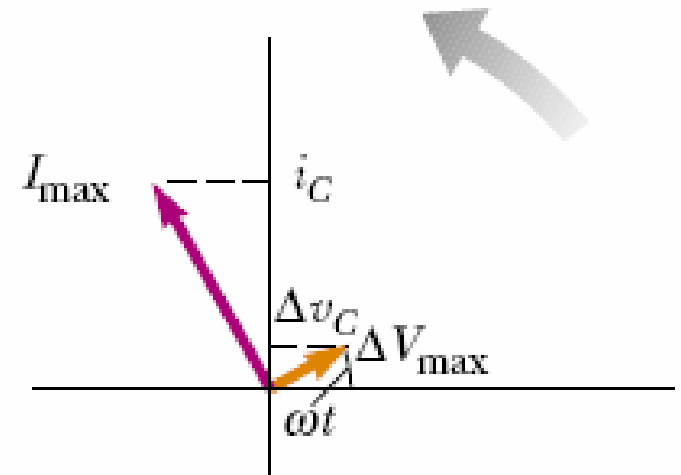
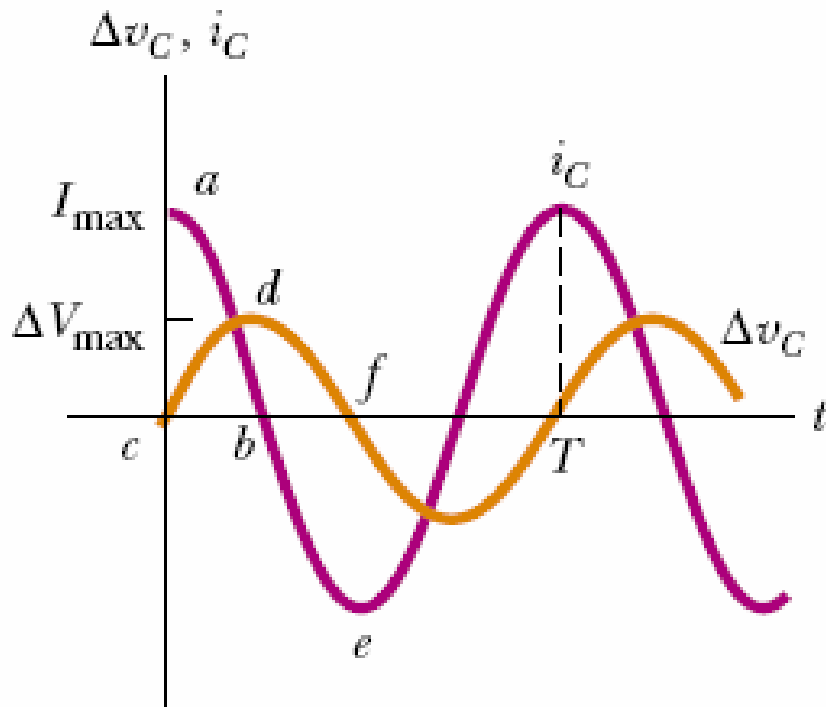
$$q = C \Delta V_{\max} \sin \omega t$$

$$i_C = \frac{dq}{dt} = \omega C \Delta V_{\max} \cos \omega t$$

Usando:

$$\cos \omega t = \sin \left( \omega t + \frac{\pi}{2} \right)$$

$$i_C = \omega C \Delta V_{\max} \sin \left( \omega t + \frac{\pi}{2} \right)$$



La corriente está **adelantada** respecto al voltaje en  $\pi/2$

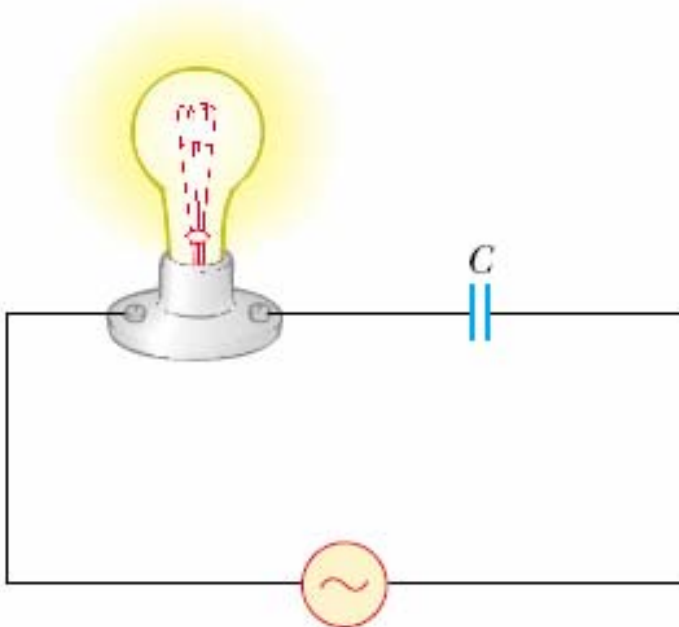
$$I_{\max} = \omega C \Delta V_{\max} = \frac{\Delta V_{\max}}{(1/\omega C)}$$

$$X_C \equiv \frac{1}{\omega C}$$

**reactancia  
capacitiva**

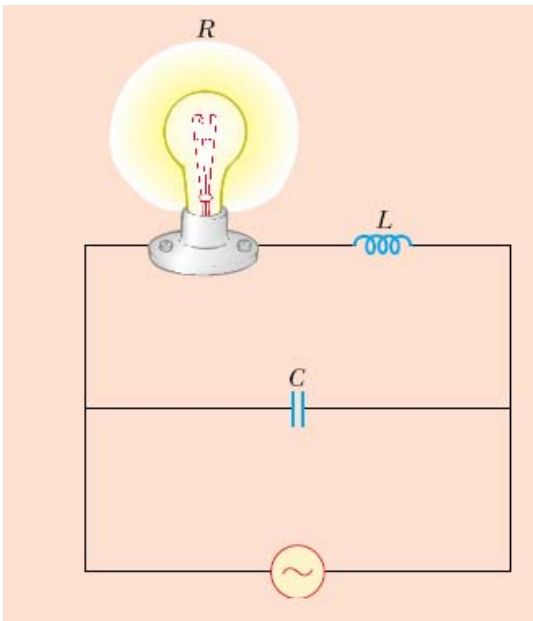
$$I_{\max} = \frac{\Delta V_{\max}}{X_C}$$

$$[X_L] = \Omega$$



**Si la frecuencia de la fuente es variable y la amplitud de V constante. La lámpara brilla más intensamente a:**

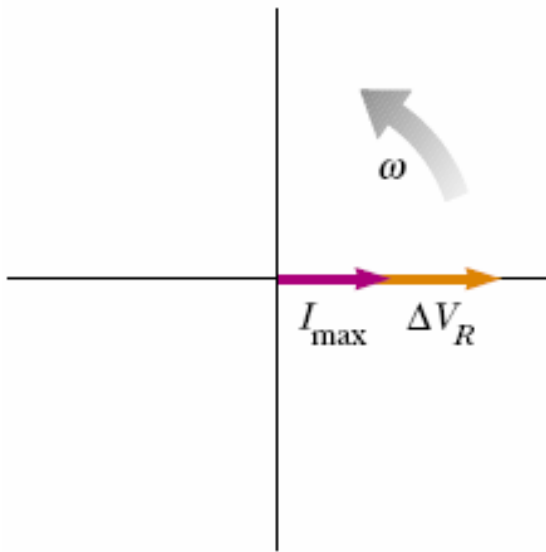
- i) altas frecuencias**
- ii) bajas frecuencias**
- iii) igual para todas**



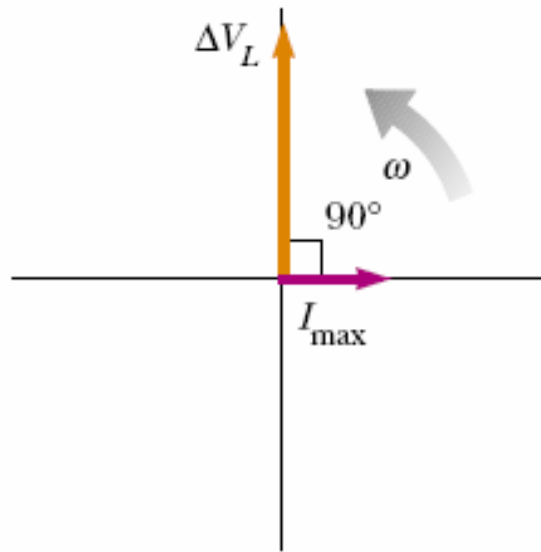
La lámpara brilla más intensamente a:

- i) altas frecuencias
- ii) bajas frecuencias
- iii) igual para todas

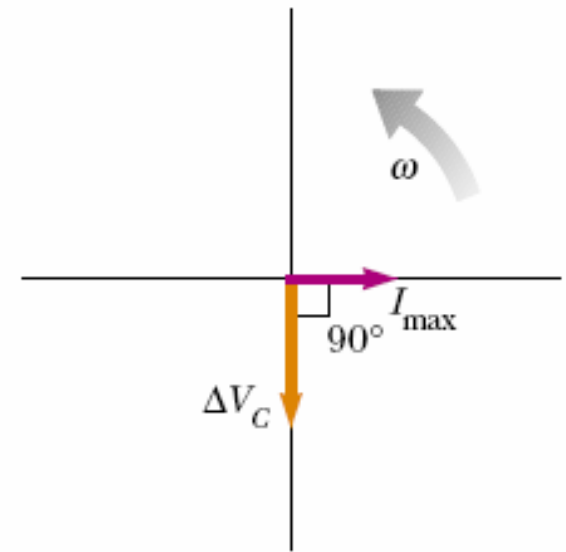
**Resumiendo:**



(a) Resistor

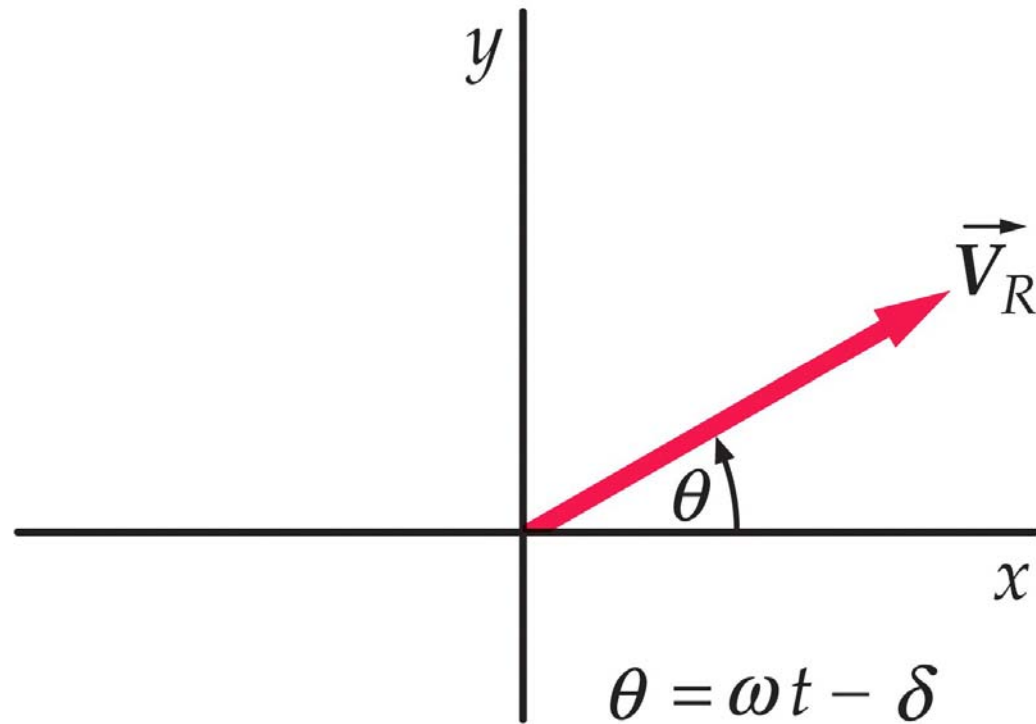


(b) Inductor



(c) Capacitor

## Representación compleja

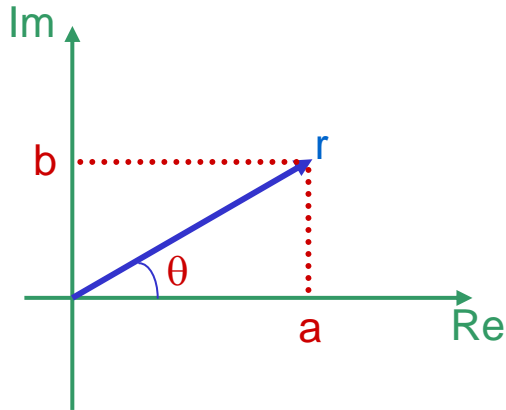


**La representación fasorial, la podemos llevar a cabo en el plano complejo:**

$y \longrightarrow \text{Im}$

$x \longrightarrow \text{Re}$





**Coordenadas cartesianas**

$$z = a + jb$$

**Coordenadas polares**

$$z = r \angle \theta$$

Cambio de coordenadas

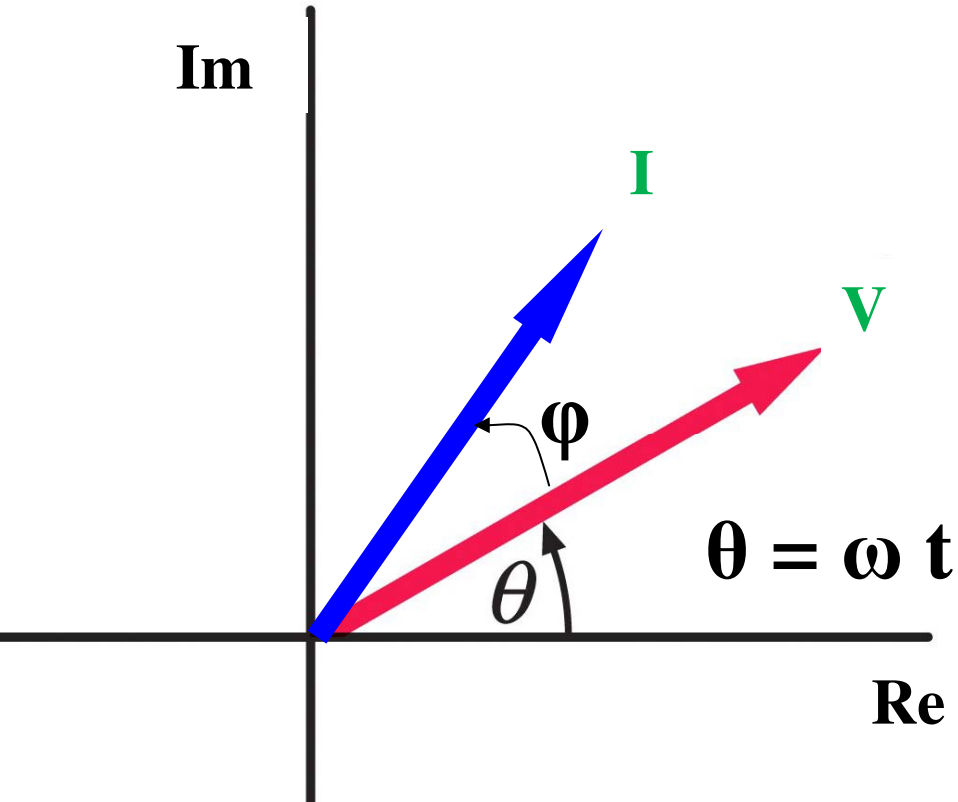
Cartesianas a polares

$$\left\{ \begin{array}{l} r = \sqrt{a^2 + b^2} \\ \theta = \arctan \frac{b}{a} \end{array} \right.$$

Polares a cartesianas

$$\left\{ \begin{array}{l} a = r \cos \theta \\ b = r \sin \theta \end{array} \right.$$

**Fórmula de Euler**  $\Rightarrow re^{\pm j\theta} = r(\cos \theta \pm j \sin \theta)$



$$\mathbf{V} = V_0 e^{j\omega t}$$

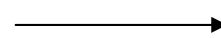
$$\mathbf{I} = I_0 e^{j\omega t + \phi}$$

**Sentido físico:**

$$V(t) = \text{Im } \mathbf{V} = V_0 \text{ sen}(\omega t)$$

$$I(t) = \text{Im } \mathbf{I} = I_0 \text{ sen}(\omega t + \phi)$$

Se opera con números complejos

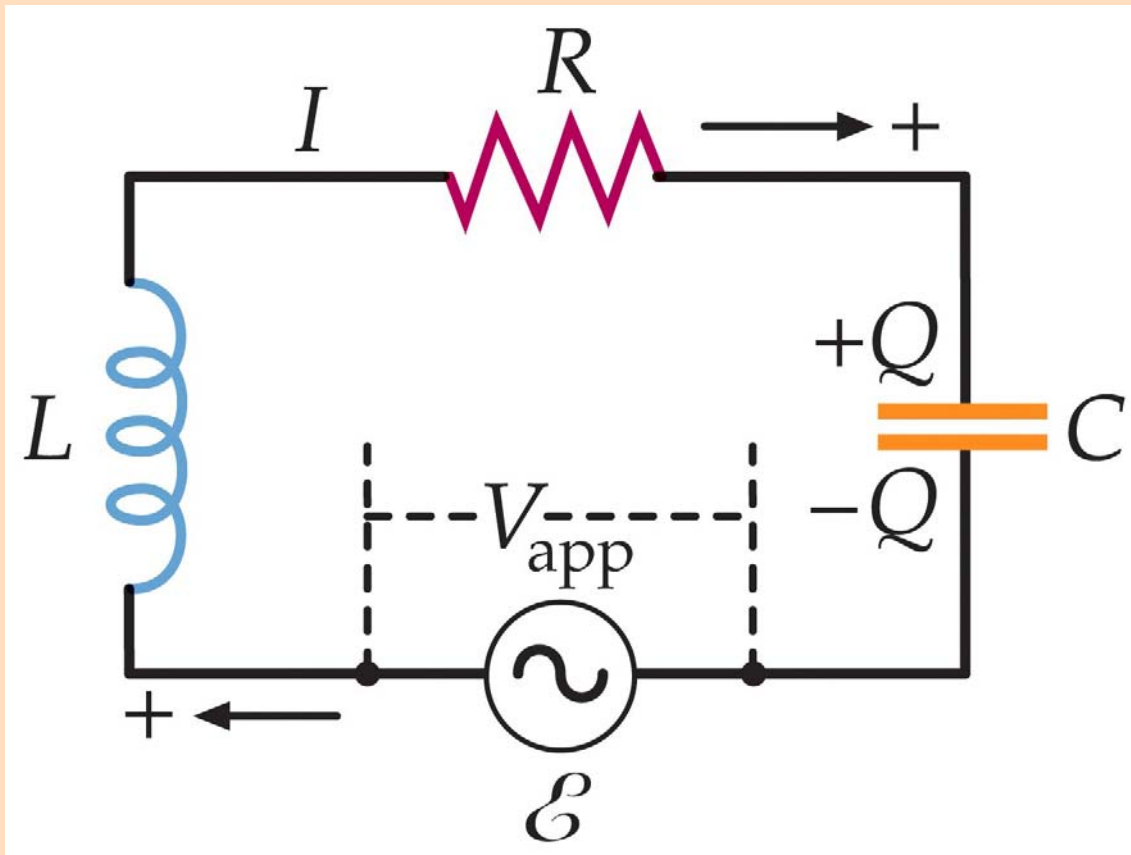


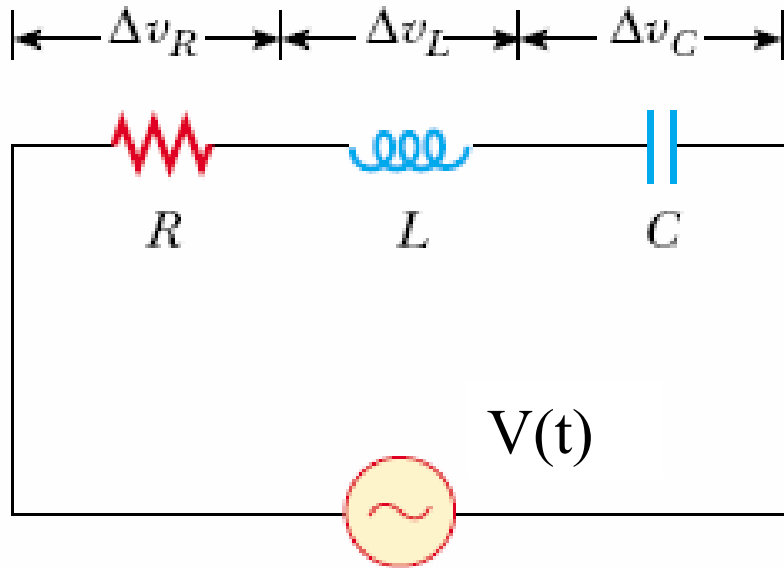
la parte Im

$$d\mathbf{I}/dt = d/dt (I_0 e^{j\omega t + \phi}) = j\omega \mathbf{I}$$

$$\int \mathbf{I} dt = \int I_0 e^{j\omega t + \phi} dt = \mathbf{I} / j\omega$$

# Circuito en serie RLC





$$V = \Delta v_R + \Delta v_L + \Delta v_C$$

$$V = I R + L \frac{dI}{dt} + q/C$$

$$V = I R + L \frac{dI}{dt} + 1/C \int I dt$$

Pasando a complejos

$$\mathbf{V} = \mathbf{I} R + L \frac{d\mathbf{I}}{dt} + 1/C \int \mathbf{I} dt$$

$$\mathbf{V} = \mathbf{I} R + j\omega L \mathbf{I} + (1/j\omega C) \mathbf{I}$$

$$\mathbf{V} = [ R + j (\omega L - 1/\omega C) ] \mathbf{I}$$

$$\mathbf{V} = [ R + j (\omega L - 1/\omega C) ] \mathbf{I}$$

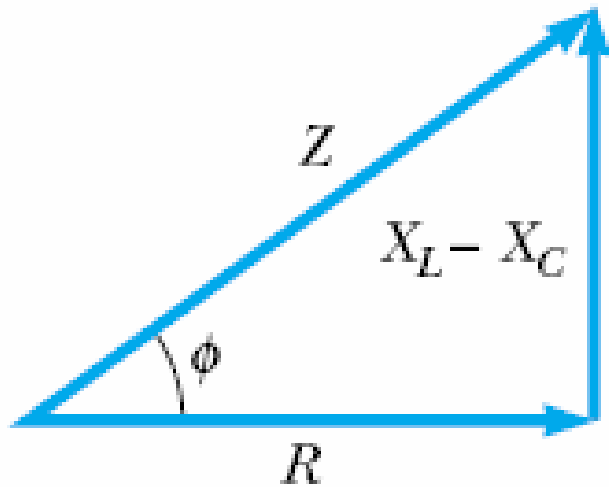
$$\mathbf{V} = [ R + j (X_L - X_C) ] \mathbf{I}$$

$$\mathbf{V} = \mathbf{Z} \mathbf{I}$$

donde

$$\mathbf{Z} = R + j (X_L - X_C)$$

**Impedancia**



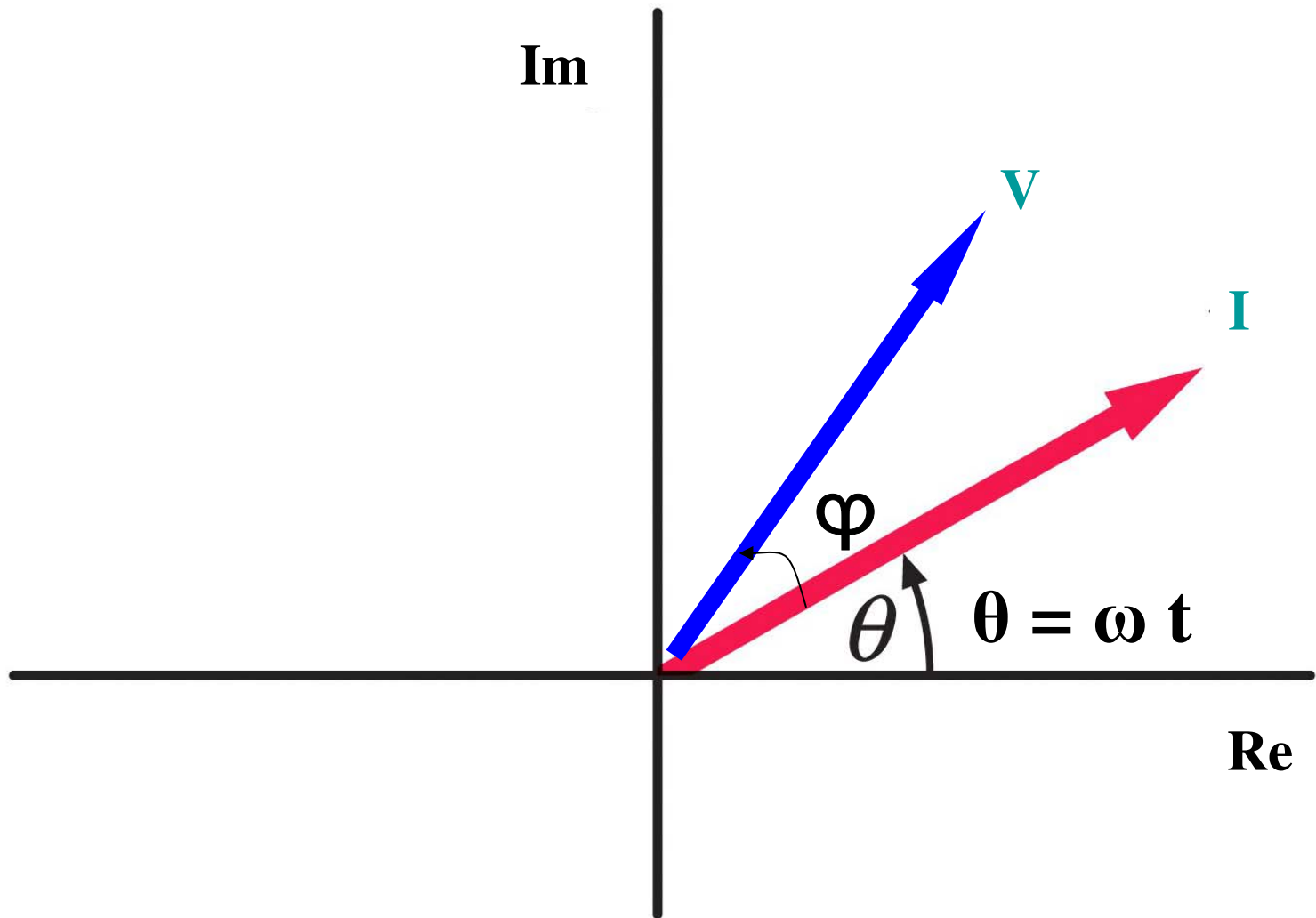
$$Z \equiv \sqrt{R^2 + (X_L - X_C)^2}$$

$$\phi = \tan^{-1} \left( \frac{X_L - X_C}{R} \right)$$

$$\mathbf{V} = \mathbf{I} \mathbf{Z} = (I_0 e^{j\omega t}) (Z e^{j\varphi}) = I_0 Z e^{j(\omega t + \varphi)}$$

$$V = V_0 \sin(\omega t + \varphi) \quad \text{donde } V_0 = I_0 Z$$

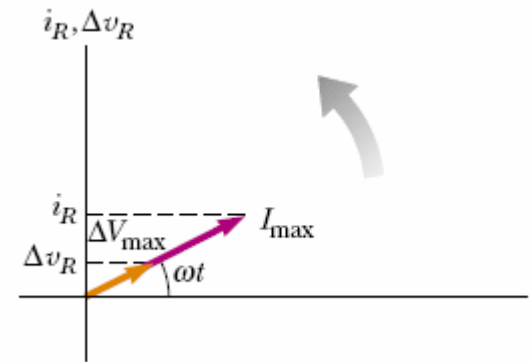
$\varphi$  : Desfasaje entre  $V$  e  $I$



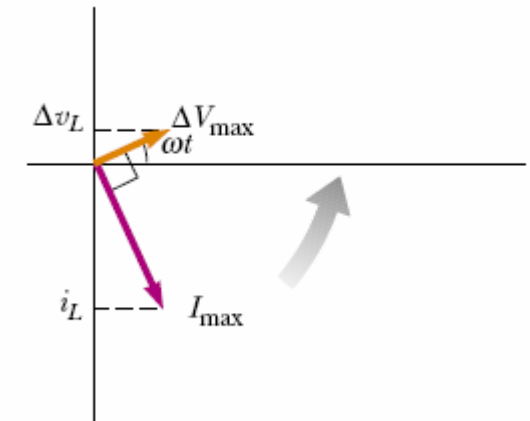
$$I_{\max} = \frac{\Delta V_{\max}}{\sqrt{R^2 + (X_L - X_C)^2}}$$

## Casos anteriores:

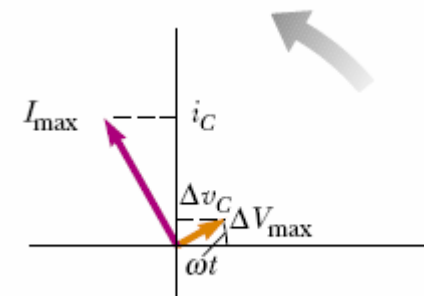
a) R puro  $\longrightarrow Z = R \quad \varphi = 0$



b) L puro  $\longrightarrow Z = \omega L \quad \varphi = \pi/2$



c) C puro  $\longrightarrow Z = 1/\omega C \quad \varphi = -\pi/2$



$$Z \equiv \sqrt{R^2 + (X_L - X_C)^2}$$



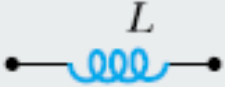


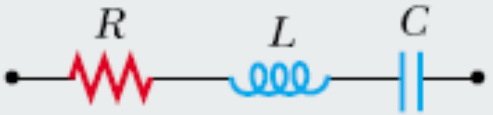
$$\phi = \tan^{-1} \left( \frac{X_L - X_C}{R} \right)$$

**Notemos que  $Z(\omega)$  y  $\varphi(\omega)$**

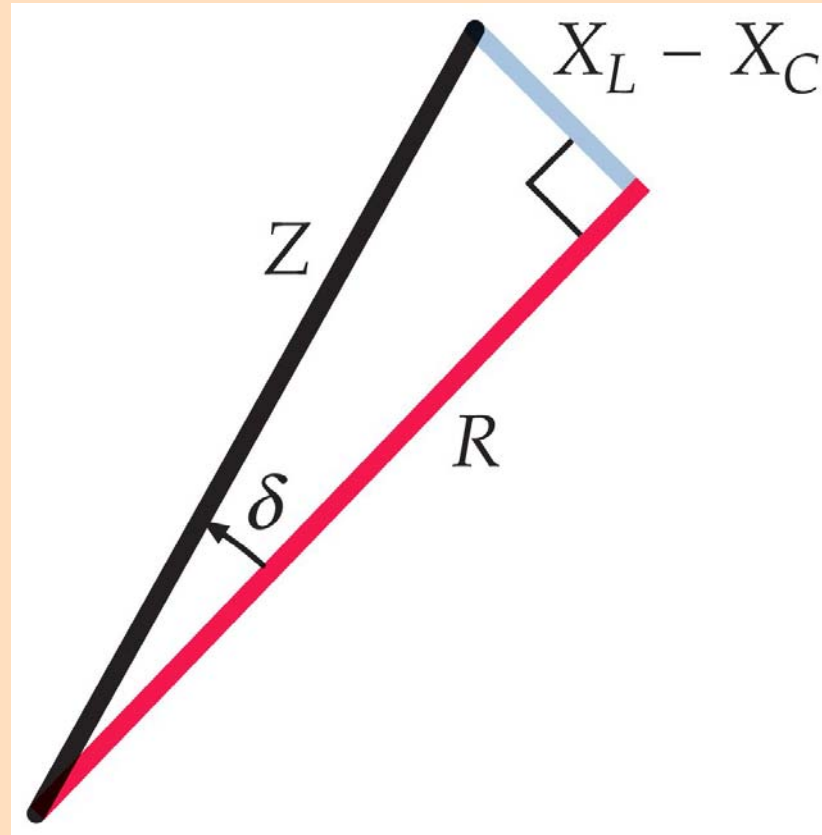
<b>Alta <math>\omega</math></b>	<b><math>X_L &gt; X_C</math></b>	<b><math>\varphi &gt; 0</math></b>	<b>I retrasada</b>
<b>Baja <math>\omega</math></b>	<b><math>X_L &lt; X_C</math></b>	<b><math>\varphi &lt; 0</math></b>	<b>I adelantada</b>



## Impedance Values and Phase Angles for Various Circuit-Element Combinations<sup>a</sup>

Circuit Elements	Impedance $Z$	Phase Angle $\phi$
	$R$	$0^\circ$
	$X_C$	$-90^\circ$
	$X_L$	$+90^\circ$
	$\sqrt{R^2 + X_C^2}$	Negative, between $-90^\circ$ and $0^\circ$
	$\sqrt{R^2 + X_L^2}$	Positive, between $0^\circ$ and $90^\circ$
	$\sqrt{R^2 + (X_L - X_C)^2}$	Negative if $X_C > X_L$ Positive if $X_C < X_L$

# Potencia en un circuito CA



**La potencia instantánea entregada por el generador es:**

$$\begin{aligned}\mathcal{P} &= i \Delta v = I_{\max} \sin(\omega t - \phi) \Delta V_{\max} \sin \omega t \\ &= I_{\max} \Delta V_{\max} \sin \omega t \sin(\omega t - \phi)\end{aligned}$$

**usando:**  $\sin(\omega t - \phi) = \sin \omega t \cos \phi - \cos \omega t \sin \phi.$

$$\mathcal{P} = I_{\max} \Delta V_{\max} \sin^2 \omega t \cos \phi - I_{\max} \Delta V_{\max} \sin \omega t \cos \omega t \sin \phi$$

**La potencia media entregada por el generador es:**

$$\mathcal{P}_{\text{av}} = \frac{1}{2} I_{\max} \Delta V_{\max} \cos \phi \longrightarrow \mathcal{P}_{\text{av}} = I_{\text{rms}} \Delta V_{\text{rms}} \cos \phi$$

$$\cos \phi = \frac{R}{Z} \quad \text{factor de potencia}$$

$$\mathbf{R \text{ puro}} \longrightarrow \cos \phi = \mathbf{1} \qquad \mathcal{P}_{\text{av}} = I_{\text{rms}} \Delta V_{\text{rms}}$$

$$\mathbf{L \text{ puro}} \longrightarrow \cos \phi = \mathbf{0} \qquad \mathcal{P}_{\text{av}} = \mathbf{0}$$

**C puro**

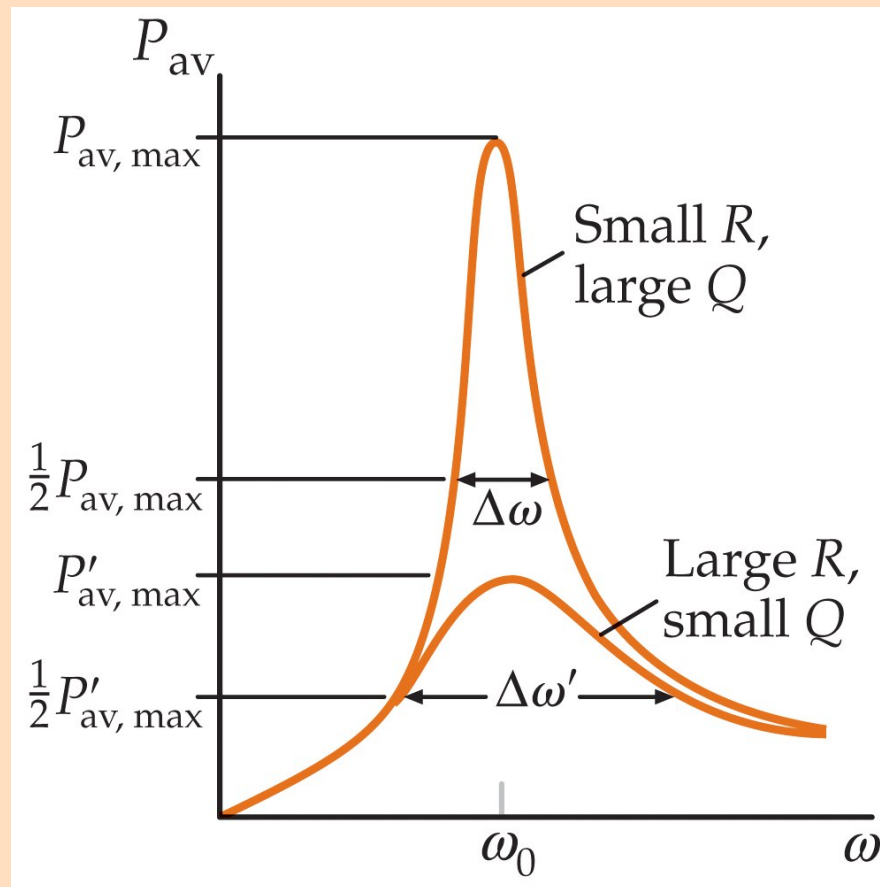
$$\cos \phi = \frac{R}{Z} \qquad \longrightarrow \qquad \cos \phi = I_{\text{max}} R / \Delta V_{\text{max}}$$
$$I_{\text{max}} = \frac{\Delta V_{\text{max}}}{\sqrt{R^2 + (X_L - X_C)^2}}$$

$$\mathcal{P}_{\text{av}} = I_{\text{rms}} \Delta V_{\text{rms}} \cos \phi = I_{\text{rms}} \left( \frac{\Delta V_{\text{max}}}{\sqrt{2}} \right) \frac{I_{\text{max}} R}{\Delta V_{\text{max}}} = I_{\text{rms}} \frac{I_{\text{max}} R}{\sqrt{2}}$$

$$\mathcal{P}_{\text{av}} = I_{\text{rms}}^2 R$$

**La potencia media entregada por el generador se disipa como calor en el resistor**

# Resonancia en un circuito RLC



**Un circuito RLC se dice que esta en resonancia cuando la corriente es máxima.**

$$I_{\text{rms}} = \frac{\Delta V_{\text{rms}}}{Z}$$

$$I_{\text{rms}} = \frac{\Delta V_{\text{rms}}}{\sqrt{R^2 + (X_L - X_C)^2}}$$

$$X_L = \omega L, X_C = 1/\omega C,$$

$$(X_L - X_C)^2 = \left(\omega L - \frac{1}{\omega C}\right)^2 = \frac{L^2}{\omega^2} (\omega^2 - \omega_0^2)^2$$

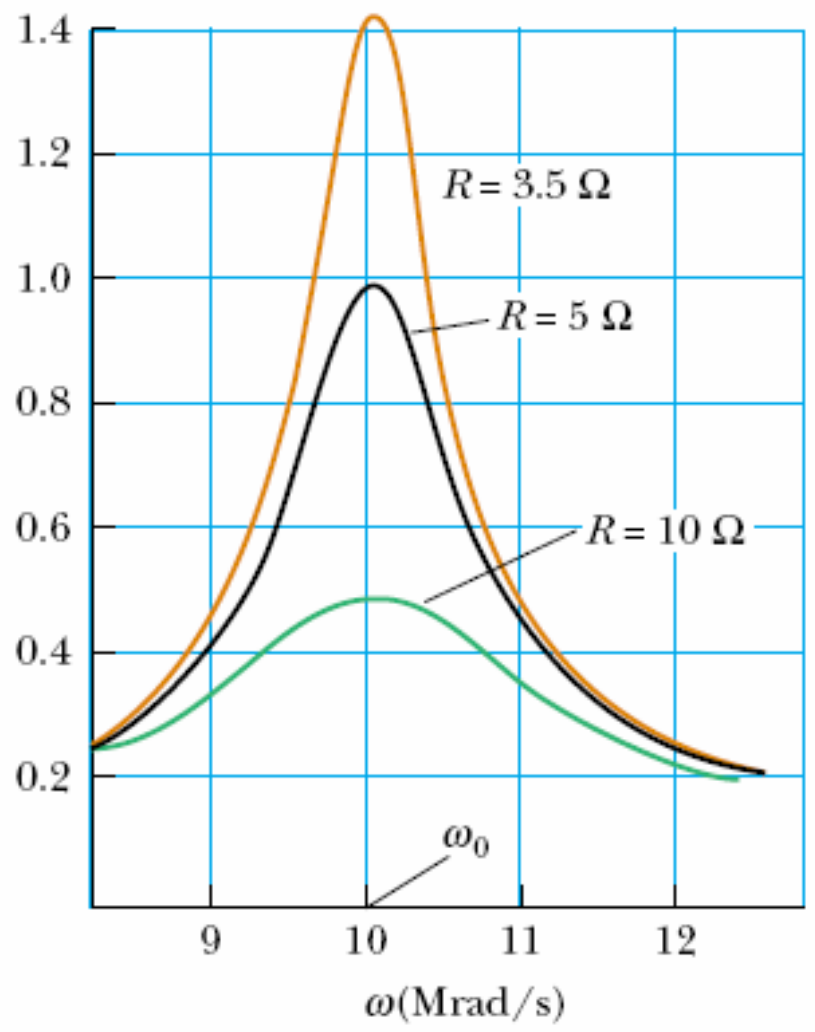
donde

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

**frecuencia de resonancia del circuito**

$L = 5.0 \mu\text{H}$   
 $C = 2.0 \text{ nF}$   
 $\Delta V_{\text{rms}} = 5.0 \text{ mV}$   
 $\omega_0 = 1.0 \times 10^7 \text{ rad/s}$

$I_{\text{rms}}$  (mA)



## Aplicación



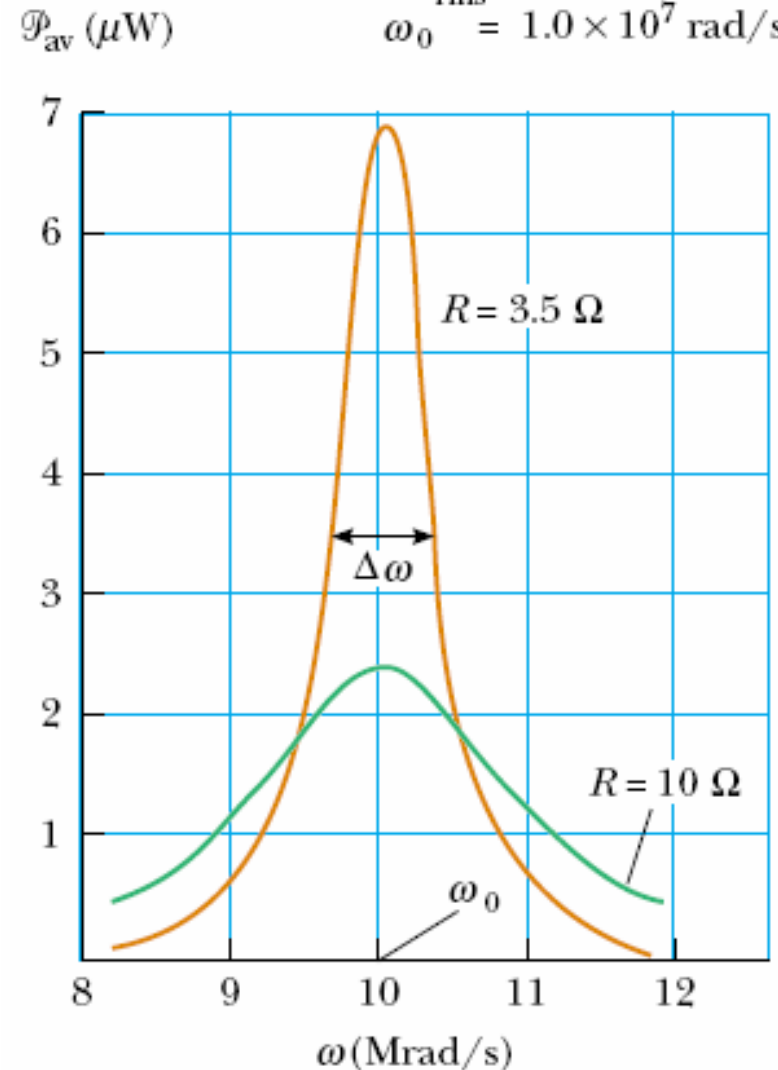
$$\mathcal{P}_{\text{av}} = I_{\text{rms}}^2 R = \frac{(\Delta V_{\text{rms}})^2}{Z^2} R = \frac{(\Delta V_{\text{rms}})^2 R}{R^2 + (X_L - X_C)^2}$$

$$\begin{aligned} L &= 5.0 \mu\text{H} \\ C &= 2.0 \text{nF} \\ \Delta V_{\text{rms}} &= 5.0 \text{mV} \\ \omega_0 &= 1.0 \times 10^7 \text{rad/s} \end{aligned}$$

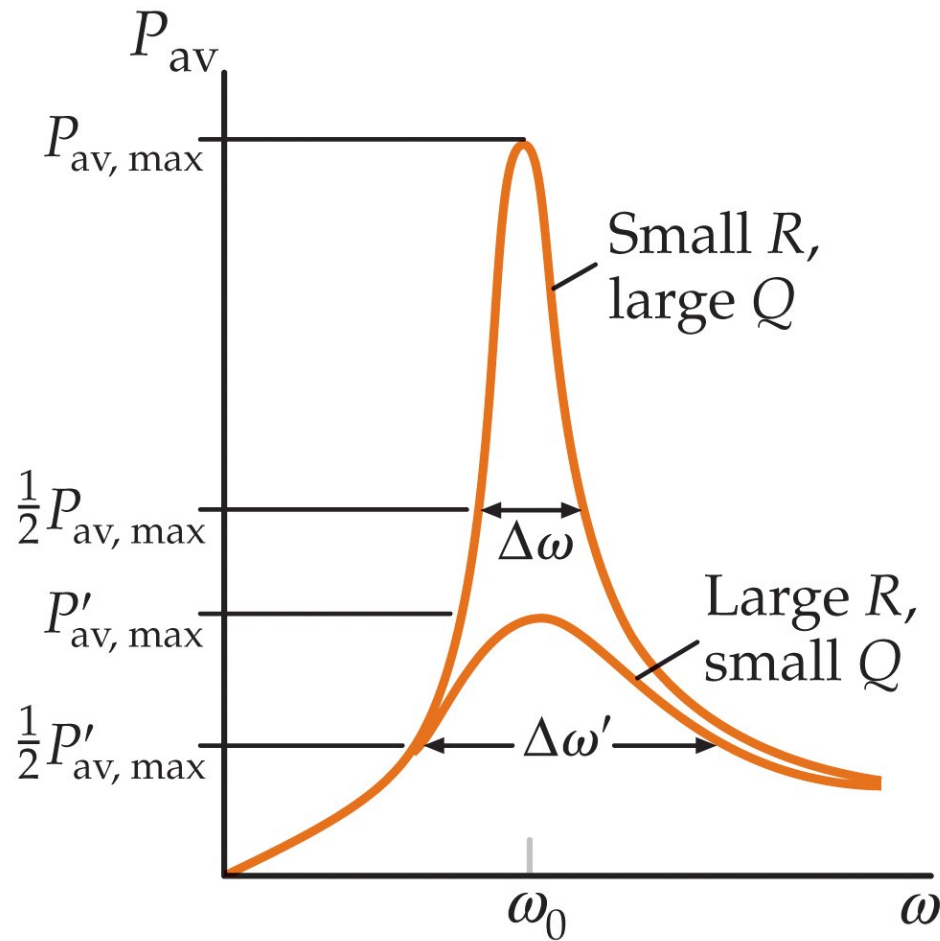
$$(X_L - X_C)^2 = \left( \omega L - \frac{1}{\omega C} \right)^2 = \frac{L^2}{\omega^2} (\omega^2 - \omega_0^2)^2$$

$$\mathcal{P}_{\text{av}} = \frac{(\Delta V_{\text{rms}})^2 R \omega^2}{R^2 \omega^2 + L^2 (\omega^2 - \omega_0^2)^2}$$

La potencia es máxima para  $\omega = \omega_0$







$$Q = \frac{\omega_0}{\Delta\omega}$$

**Factor de calidad o de mérito**

**Se puede probar que en un circuito RLC:**

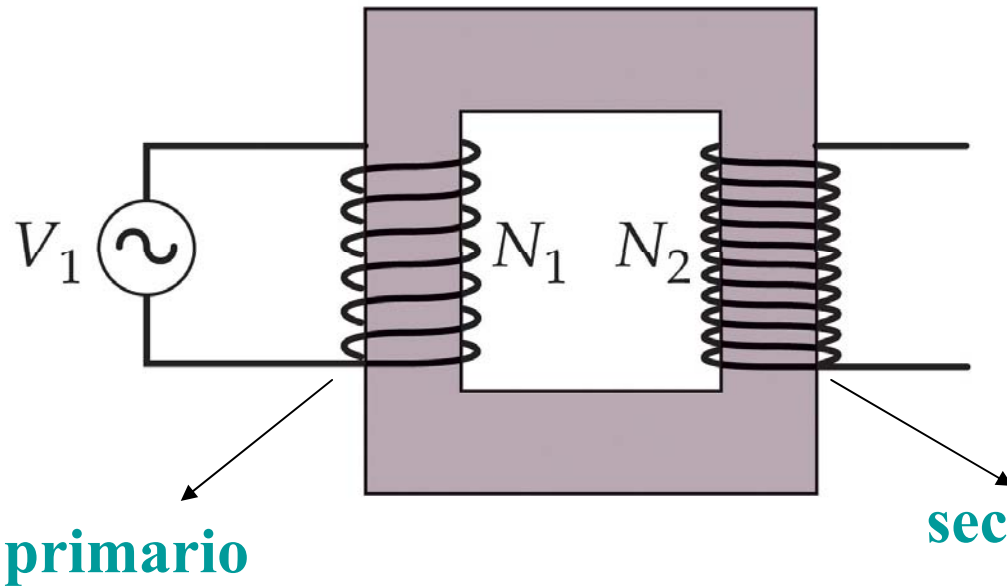
$$\Delta\omega = R/L$$

$$Q = \frac{\omega_0 L}{R}$$

**Valores típicos de  $Q$ : 10-100**

# Transformador





$$\Delta V_1 = -N_1 \frac{d\Phi_B}{dt}$$

$$\Delta V_2 = -N_2 \frac{d\Phi_B}{dt}$$

Si suponemos que no hay pérdidas de flujo fuera del núcleo de hierro

$$\longrightarrow \Delta V_2 = \frac{N_2}{N_1} \Delta V_1$$

Dependiendo de  $N_1$  y  $N_2$ , podemos tener un **elevador** o un **reductor** de voltaje

**Cerrando el circuito secundario y admitiendo pérdidas de energía por unidad de tiempo pequeñas, la potencia entregada por el primario será igual a la del secundario**

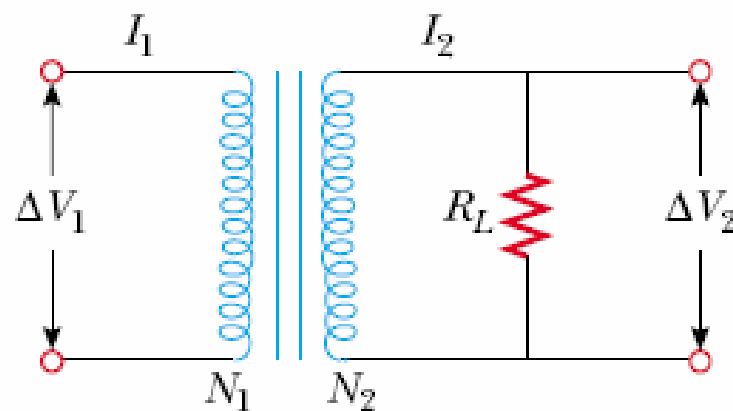
$$I_1 \Delta V_1 = I_2 \Delta V_2 \quad \longrightarrow \quad I_1 = \left( \frac{V_2}{V_1} \right) I_2 = \left( \frac{N_2}{N_1} \right) I_2$$

**Transformador reductor**

**V** ↓  
**I** ↑

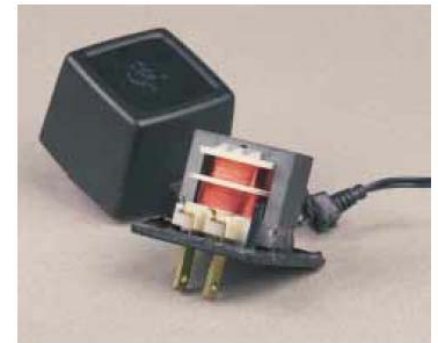
**Transformador elevador**

**V** ↑  
**I** ↓



**símbolo**

**Los núcleos de hierro se laminan para evitar pérdidas por corrientes parasitas**



# Guerra de las corrientes : CA vs. CC



**George Westinghouse  
(1846-1914)**

**En 1886 fundó Westinghouse  
Electric**

**Vs.**

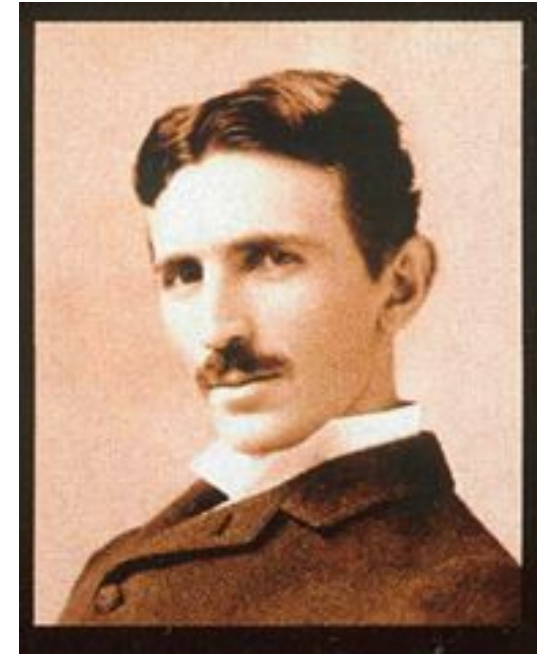
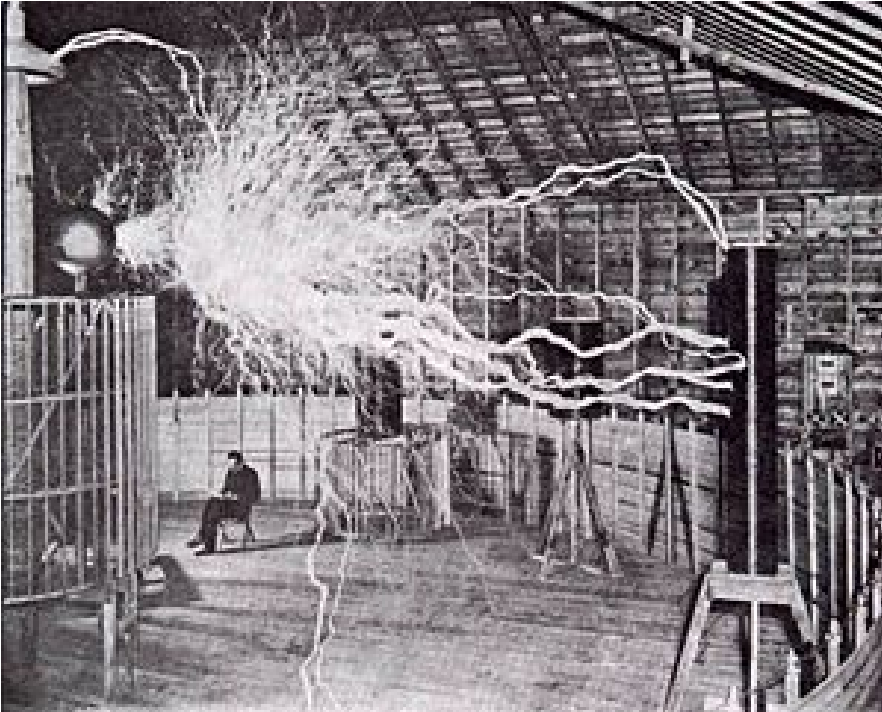


**Thomas Alva Edison  
(1847-1931)**

**En 1880 se asocia con J.P. Morgan  
para fundar la General Electric**



## **Nikola Tesla (1856-1943)**



**Con el apoyo financiero de George Westinghouse, la corriente alterna sustituyó a la continua. Tesla fue considerado desde entonces el fundador de la industria eléctrica.**



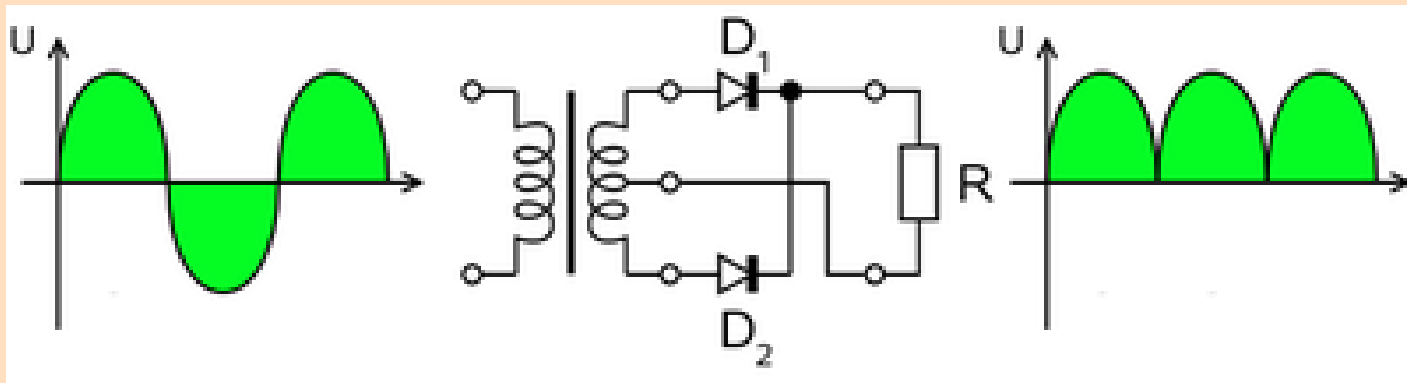
**Bobina de Tesla:** están compuestas por una serie de circuitos eléctricos resonantes acoplados. Crean descargas eléctricas de largo alcance.

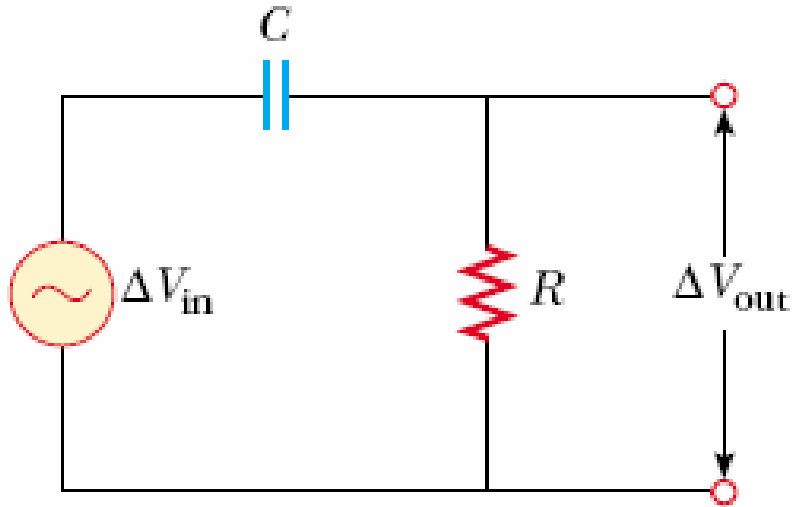


**Torre Tesla:** torre-antena de telecomunicaciones inalámbricas pionera diseñada para demostrar la transmisión de energía sin cables conectores entre los años 1901 y 1917.



# Rectificadores y Filtros

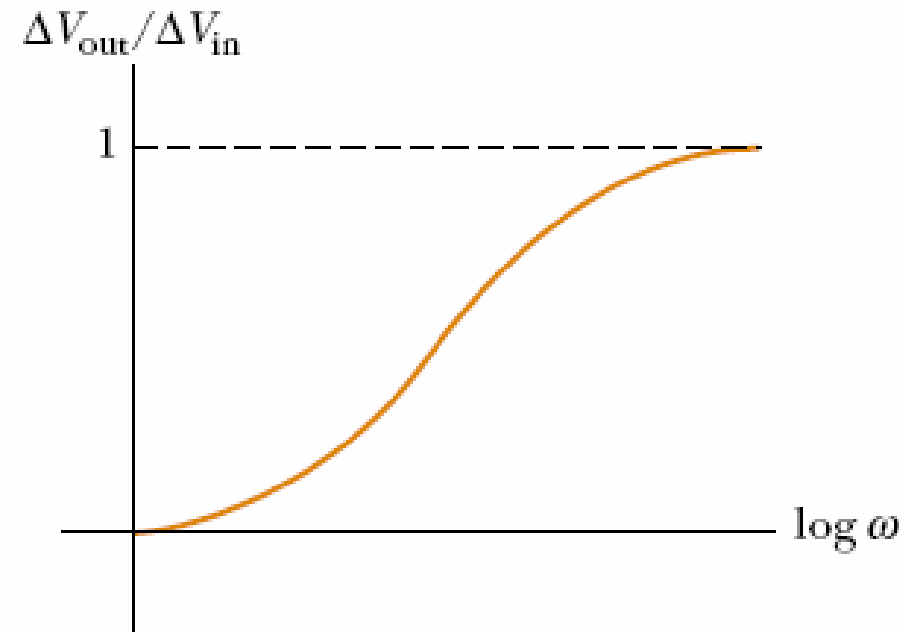




$$V_{in} = V^0_{in} \text{ sen } (\omega t)$$

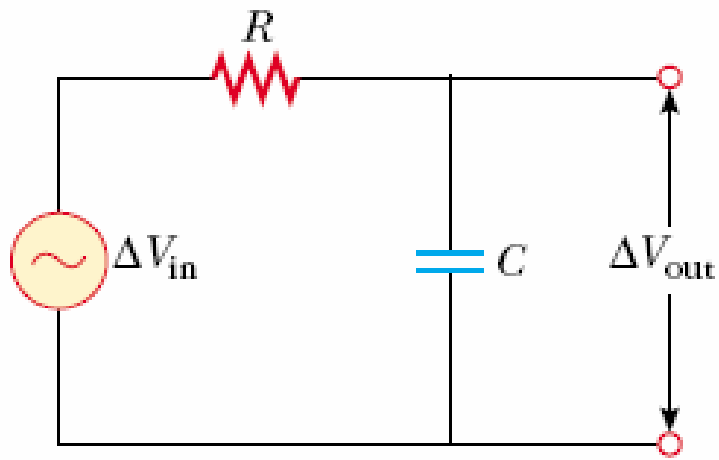
$$V^0_{in} = I^0 Z = I^0 (R^2 + (1/\omega C)^2)^{1/2}$$

$$V^0_{out} = I^0 R$$



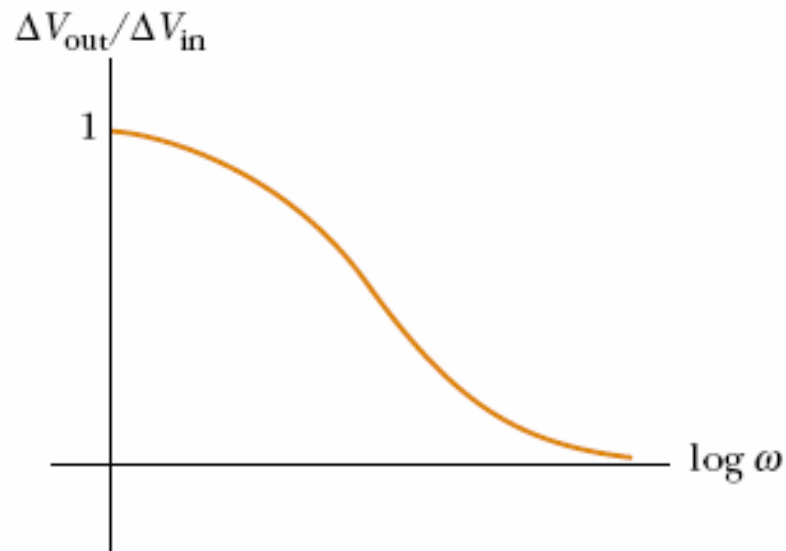
$$\frac{V^0_{out}}{V^0_{in}} = \frac{R}{(R^2 + (1/\omega C)^2)^{1/2}}$$

**Filtro pasa altos**



$$V_{in}^0 = I^0 Z = I^0 (R^2 + (1/\omega C)^2)^{1/2}$$

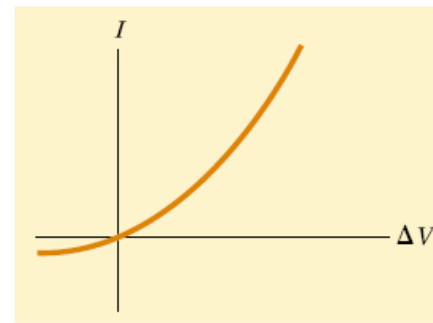
$$V_{out}^0 = I^0 X_c = I^0 / \omega C$$



$$\frac{V_{out}^0}{V_{in}^0} = \frac{1/\omega C}{(R^2 + (1/\omega C)^2)^{1/2}}$$

**Filtro pasa bajos**

# Rectificador



**Símbolo  
del iodo**

